

- **0 TO 27MHz PIXEL RATE IN SINGLE** PRECISION MODE,
- **0 TO 20 MHz PIXEL RATE IN DOUBLE** PRECISION MODE
- **FORWARD AND INVERSE 8 x 8 TRANS-**FORM ■ FORWARD AND INVERSE 8 x 8 TRANS-
FORM
■ 9-BIT TWO'S COMPLEMENT PIXEL FORMAT

Ayy SGS-THOMSON
MICROELECTRONICS

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- FORM
■ 9-BIT TWO'S COMPLEMENT PIXEL FORMAT
■ 12-BIT TWO'S COMPLEMENT COEFFICIENT FORMAT ■ 12-BIT TWO'S COMPLEMENT COEFFICIENT
FORMAT
OPTIMIZED ACCURACY FOR 8-BIT TWO'S
- COMPLEMENT PIXEL FORMAT
- **E. SELECTABLE SCANNING OF COEFFICIENT BLOCKS** ■ SELECTABLE SCANNING OF COEFF
BLOCKS
■ FULLY TTL AND CMOS COMPATIBLE
-
- BLOCKS
■ FULLY TTL AND CMOS
■ CMOS TECHNOLOGY
- CMOS TECHNOLOGY
■ SINGLE +5 VOLT POWER SUPPLY
- **MAXIMUM POWER DISSIPATION : 750mW** AT 27MHz

DESCRIPTION

The STV3208 is a dedicated circuit for the 8 x 8 discrete cosine transform (DCT) computation. Twodimensional forward DCT (FDCT) or inverse DCT (IDCT) is performedfor 8 x 8 block sizes and a pixel rate up to 27MHz. The circuit architecture is fully bidirectional with 9-bit magnitude pixel data bus and a 12-bit magnitude coefficient data bus programmed as input or output depending on the selection of FDCT or IDCT.

For the forward transform, the input pixels are coded on 9-bit 2's complement and the output coefficients are coded on 12-bit 2's complement. For the inverse transform, the data format is identical with the coefficients used as input and the pixels used as output.

Two operating modes are provided : single precision modeat a pixel rate up to 27 MHz, and double precision mode at a pixel rate up to 20 MHz.

ORDER CODES

PIN CONNECTIONS

PIN IDENTIFICATION

FUNCTIONAL BLOCK DIAGRAM

FUNCTIONAL DESCRIPTION

1. EQUATIONS

Figure 1

The STV3208performs8 x 8 two dimensional Discrete Cosine Transformaccordingto thefollowingformula:

Equations for 9-bit PIXEL DATA (PR pin set to low) :

FORWARD TRANSFORM EQUATION :

$$
F(u, v) = \text{Round}\left[\frac{1}{4}C(u) C(v) \sum_{i=0}^{7} \sum_{j=0}^{7} D(i, j) \cos\frac{(2 \cdot i + 1) u \pi}{16} \cos\frac{(2 \cdot j + 1) v \pi}{16}\right]
$$

INVERSE TRANSFORM EQUATION :

D(i, j) = Round
$$
\left[\frac{1}{4} \sum_{u=0}^{7} \sum_{v=0}^{7} C(u) C(v) F(u, v) \cos \frac{(2 \cdot i + 1) u \pi}{16} \cos \frac{(2 \cdot j + 1) v \pi}{16} \right]
$$

Where C(u) = $\frac{1}{\sqrt{2}}$ if u = 0

= 1 otherwise

Equations for 8-bit PIXEL DATA (PR pin set to high) :

FORWARD TRANSFORM EQUATION :

$$
F(u, v) = \text{Round} \left[\frac{1}{2} C(u) C(v) \sum_{i=0}^{7} \sum_{j=0}^{7} D(i, j) \cos \frac{(2 \cdot i + 1) u \pi}{16} \cos \frac{(2 \cdot j + 1) v \pi}{16} \right]
$$
\n
$$
INVERSE TRANSFORM EQUATION:
$$
\n
$$
D(i, j) = \text{Round} \left[\frac{1}{8} \sum_{u=0}^{7} \sum_{v=0}^{7} C(u) C(v) F(u, v) \cos \frac{(2 \cdot i + 1) u \pi}{16} \cos \frac{(2 \cdot j + 1) v \pi}{16} \right]
$$

Where C(u) = $\frac{1}{\sqrt{2}}$ if u = 0 = 1 otherwise

2. DATA FLOW ORDERING

The pixel block is scanned column by column (ORDER 1) or line by line (ORDER 2). If CSS is high, the coefficient block is scanned with a zig zag order. Figure 2 shows the relation between pixels order and coefficient order.

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Figure 2a : Data Ordering (CSS high)

If CSS is low, the coefficient block is scanned line by line and the pixel block is scanned column by column, or the coefficient block is scanned column by column and the pixel block is scanned line by line.

Figure 2b : Data Ordering (CSS low)

3. DATA FORMAT

Coefficients format is 12-bit 2's complement, corresponding to the range -2048 to 2047.

There are 2 possible ranges for pixel data :

9-bit two's complement magnitude (see Figure 3)

The pixel data range is -256 to +255. In this case the PRpin must be set to 0 for IDCT. D8 is the most significant bit and D0 the least significant bit for the pixel data. A clipping to the range -256 to +255 is performed before outputting reconstructed pixels after an IDCT.

8-bit two's complement magnitude

(see Figure 4)

Figure 3

Pixel data range is -128 to +127. In this case D0 must be set to 0 and the PR Pin must be set to 1 for IDCT. D8 is the most significant bit and D1 the least significant bit for the pixel data. A clipping to the range -128 to +127 is performed before outputting reconstructed pixels after an IDCT.

This mode may be used for intra picture coding. In this case, pixel data range is 0 to 255. For a FDCT, the most significant bit of input pixel data (D8) must be invertedbefore entering the chip. This isequivalent to substract 128 to the input pixel data. Note that this operation will only have effect on the DC value F(0,0). For an IDCT, the most significant bit of output pixel data (D8) must be inverted. This is equivalent to add 128 to the output pixel data.

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Figure 4

4. BLOCK FLOW

Depending on the application, blocks may be entered in different way.

Latent period :

The latent period between input data and corresponding output results is 167 clock cycles (if FDCT is selected)or 163 clock cycles (if IDCT is selected) in single precision mode (S/D pin set to 1). This means that the first data of the resulting block is provided 137 clock cycles (if FDCT is selected) or 135 clock cycles (if IDCT is selected) in double precision mode (S/D pin set

to 0).

Synchronization signals :

An input block synchronization signal must be provided. The input pin for this signal is DSYNC if FDCT is selected and FSYNC if IDCT is selected.This signal is active low and must not be active more than one clock cycle and during the

Figure 5

first clock cycle after power-up.This signal must be active with the first data of each input block or group of blocks.

An output block synchronizationsignal is provided. The output pin for this signal is FSYNC if FDCT is selected and DSYNC if IDCT is selected. This signal is active with the first data of each output block or group of blocks.

The output synchronization signal is equal to the input synchronization signal delayed from the latent period (see Figure 5).

CONTINUOUS BLOCK FLOW

Inputs data are fed continouslywith one new item data at each clock cycle and output data is provided continously with one new result data item at each clock cycle.

The input synchronization signal can be provided for each input block. In this case the output synchronizationpulse isprovided for eachoutput block (Figure 6). An other way is to provide a synchronization pulse only for the first block of a group of blocks. In this case, only one synchronizationpulse is provided for the first output block (Figure 7).

Figure 7 : Continuous Block Flow 2

CONTINUOUS BLOCK FLOW WITH BYPASS OF IRREVELANT DATA

It is possible to process a block flow including irrelevant data (corresponding to line suppression for example) asif it was a continousblock flow. One way is to stop the clock signal during the irrelevant data occurence. Another way is to use the Clock Enable Signal (EN) which allows to stop the chip internal clock during irrelevant data occurrence (see Figure 8)

Figure 8 : Continuous Block Flow with Irrelevant Data

BURST BLOCK FLOW (see Figure 9)

Single blocks (or groups of block) may not be continous. In other words, delay cycles between two blocks (or groups of block) may exist. During these delay cycles, the clock is still running and the chip continues to perform computations. The constraint is that the internal pipe line must not be broken when the new block occurs.To take this constraint into account, the numberof delay cycles (NC) must respect one of the following conditions:

1 - the numberofdelay cycles (NC) isgreaterthan or equal to the latency. In this case the pipe line is empty (all the relevant data has been output) when a new input block processing starts.

2 - the number of delay cycles (NC) is a multiple of 64. In this case, the input data always remains synchronous with the internal pipe line.

MIXED FDCT/IDCT (see Figure 10)

In some low frequency application, it could be cost effective to use only one chip to compute all the DCT required by the coding scheme. Blocks must be fed in a burst fashion with at least the latency time between the last pixel of input pixels for FDCT and the first pixel of input coefficients for IDCT. The same delay must be respected between the last pixel of input coefficientsfor IDCT and the first pixel of input pixels for FDCT.

Figure 10 : Mixed 8 x 8 FDCT/IDCT Example Waveforms (double precision)

PRECISION SELECTION

For single precision mode, the S/D pin must be high. In this case, the maximum rating for pixel is 27 MHz. For double precision mode, the S/D pin must be low. In this case, the maximum rating for pixel is 20 MHz.

5. PINS DESCRIPTION

CLK : Clock signal

DATA PINS

D0 to D8 : 9-bit bidirectional pixel data bus pins. Direction is programmed by the F/I pin :

Data is loaded (when input) or settled (when out-

put) on rising edge of CLK. D0 is the least significant bit and D8 the most significant one. Note that for the optimized mode for 8-bit 2's Cpixel data, D1 is the least significant bit and D0 must be set to 0.

DSYNC : pixel data block synchronization signal. This pin is bidirectional with the direction programmed by the F/I pin (like D0 to D8). DSYNC is active (low level during one clock cycle only) with the first pixel data of a block (or a group of blocks).

F0 to F11 : 12-bit bidirectional coefficient data bus pins. Direction is programmed by the F/I pin :

Data is loaded (when input) or settled (when output) on rising edge of CLK. F0 is the least significant bit and F11 the most significant one.

FSYNC : coefficient data block synchronization signal. This pin is bidirectional with the direction programmed by the F/I pin (like F0 to F11). FSYNC is active (low level during one clock cycle only) with the first coefficient data of a block (or a group of blocks).

CONTROL PINS

F/I : forward or inverse selection. When F/I is high, forward DCT is performed, when F/I is low, inverse DCT is performed.

S/D : single or double precision. When S/D is high, single precision is selected. When S/D is low, double precision is selected.

CSS : coefficient scanning selection. When CSS is high, zig zag scanning of coefficient block is selected. When CSS is low, row scanning of coefficient block is selected.

PR : pixel range selection. If PR is low, pixel range is -256 to +255. If PR is high, pixel range is -128 to +127.

OE : output enable. This signal is active low. When OE is high, all outputs (defined by the F/I pin state)

6. ACCURACY CHARACTERISTICS

are forced to the high impedance state.

EN : enable. This signal is active low. When EN is high, internal states of the chip are frozen. When EN becomes low, execution restarts. EN must go to high state when CLK is high.

POWER SUPPLY AND GROUND PINS

V_{cc} : +5 Volt power supply **Vss** : ground potential

OTHERS : Test. This pin is reserved and must be low in normal mode.

The accuracy characteristics have been measured according to the following scheme :

A :

Characteristics of IDCT. Error between the IDCT computed with 64-bit floating point accuracy and the IDCT computedby the STV3208 is measured. Measures have beendone according to the CCITT WGXV method

B :

Characteristics of FDCT. Error between the FDCT computed with 64-bit floating point accuracy and the FDCT computed by the STV3208 is measured.

C :

Characteristics of FDCT followed by an IDCT. Error between the source picture and the FDCT computed by the STV3208 followed by an IDCT computed by the STV3208 is measured.

TIMING WAVEFORMS

Synchonization Signals Timing Diagram for a Forward Transform

Note : FSYNC will be in unknown state after the power up during a count of cycles equal to the latency.

Synchronization Signals Timing Diagram for an Inverse Transform

Note : DSYNC will be in unknown state after the power up during a count of cycles equal to the latency.

Output Enable Signal Timing Waveforms

Control Static Signal Timing Waveforms

TIMING WAVEFORMS (continued)

Note : EN signal must change from low to high level during the high level of CLK signal.

Clock Timing Waveforms

Output Timing Waveforms

ELECTRICAL CHARACTERISTICS

ABSOLUTE MAXIMUM RATINGS Supply voltage (Vcc) : 6 Volts Operating temperature range : 0 to 70 °C

DC ELECTRICAL CHARACTERISTICS

Operating Conditions : $V_{SS} = 0$ Volt, $T_A = 0$ to 70 °C, $V_{CC} = 5$ V \pm 5% unless otherwise specified

AC ELECTRICAL CHARACTERISTICS

Operating conditions : $V_{SS} = 0$ Volt, T_A = 0 to 70 °C, V_{CC} = 5 V ± 5 % unless otherwise specified Outputs Loads : Capacitance = 50 pF, Current Logic Low = 500µA Test Load on Outputs :

Timings are measured between threshold voltage of 1.5 V unless otherwise specified.

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PACKAGE MECHANICAL DATA

40 PINS - PLASTIC DIP

PACKAGE MECHANICAL DATA

44 PINS - PLASTIC QUAD FLAT PACK

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